

# Circuits and Systems I

LECTURE #13

Frequency Response of FIR Filters and Digital Filtering of Analog Signals



Prof. Dr. Volkan Cevher

**LIONS/Laboratory for Information and Inference Systems** 

## License Info for SPFirst Slides

 This work released under a <u>Creative Commons License</u> with the following terms:

#### Attribution

■ The licensor permits others to copy, distribute, display, and perform the work. In return, licensees must give the original authors credit.

#### Non-Commercial

 The licensor permits others to copy, distribute, display, and perform the work. In return, licensees may not use the work for commercial purposes—unless they get the licensor's permission.

#### Share Alike

- The licensor permits others to distribute derivative works only under a license identical to the one that governs the licensor's work.
- Full Text of the License
- This (hidden) page should be kept with the presentation

# Outline - Today

Today

<>

Section 6-6

Section 6-7

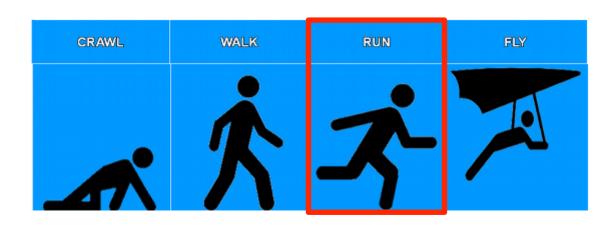
Section 6-8

Next week

<>

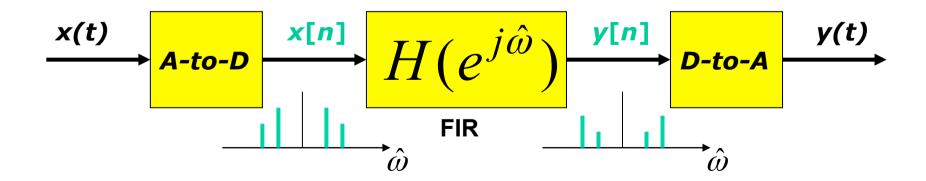
Final Exam Review

CSI
Progress
Level:



### LECTURE OBJECTIVES

- Two Domains: Time & Frequency
- Track the spectrum of x[n] thru an FIR Filter:
   Sinusoid-IN gives Sinusoid-OUT
- UNIFICATION: How does frequency response affect x(t) to produce y(t)?



# TIME & FREQUENCY

#### FIR DIFFERENCE EQUATION is the TIME-DOMAIN

$$y[n] = \sum_{k=0}^{M} b_k x[n-k] = \sum_{k=0}^{M} h[k]x[n-k]$$

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^{M} h[k]e^{-j\hat{\omega}k}$$

$$H(e^{j\hat{\omega}}) = h[0] + h[1]e^{-j\hat{\omega}} + h[2]e^{-j2\hat{\omega}} + h[3]e^{-j3\hat{\omega}} + \cdots$$

# Ex: DELAY by 2 SYSTEM

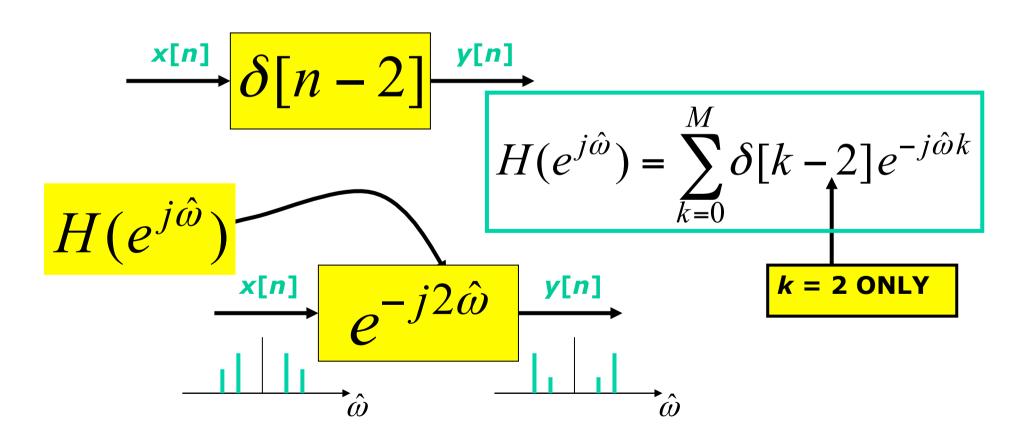
Find 
$$h[n]$$
 and  $H(e^{j\hat{\omega}})$  for  $y[n] = x[n-2]$ 

$$h[n] \qquad b_k = \{0, 0, 1\}$$

$$h[n] = \delta[n-2]$$

# DELAY by 2 SYSTEM

Find h[n] and  $H(e^{j\hat{\omega}})$  for y[n] = x[n-2]



### GENERAL DELAY PROPERTY

Find 
$$h[n]$$
 and  $H(e^{j\hat{\omega}})$  for  $y[n] = x[n - n_d]$ 

$$h[n] = \delta[n - n_d]$$

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^{M} \delta[k - n_d] e^{-j\hat{\omega}k} = e^{-j\hat{\omega}n_d}$$

ONLY ONE non-ZERO TERM for k at  $k = n_d$ 

# FREQ DOMAIN --> TIME ??

• START with  $H(e^{j\hat{\omega}})$  and find h[n] or  $b_k$  $h[n] \xrightarrow{y[n]} h[n] = ?$  $\left| H(e^{j\hat{\omega}}) = 7e^{-j2\hat{\omega}}\cos(\hat{\omega}) \right|$ 

# FREQ DOMAIN --> TIME

$$H(e^{j\hat{\omega}}) = 7e^{-j2\hat{\omega}}\cos(\hat{\omega})$$

$$= 7e^{-j2\hat{\omega}}(0.5e^{j\hat{\omega}} + 0.5e^{-j\hat{\omega}})$$

$$= (3.5e^{-j\hat{\omega}} + 3.5e^{-j3\hat{\omega}})$$

$$h[n] = 3.5\delta[n-1] + 3.5\delta[n-3]$$

$$b_k = \{0, 3.5, 0, 3.5\}$$

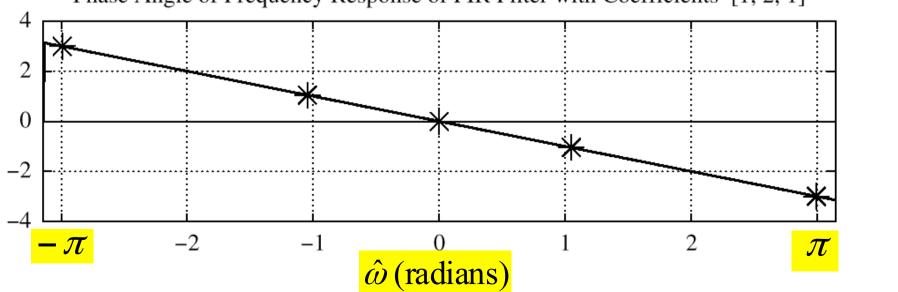
## PREVIOUS LECTURE REVIEW

- SINUSOIDAL INPUT SIGNAL
  - OUTPUT has SAME FREQUENCY
  - DIFFERENT Amplitude and Phase
- FREQUENCY RESPONSE of FIR
  - MAGNITUDE vs. Frequency
  - PHASE vs. Freq
  - PLOTTING

$$H(e^{j\hat{\omega}}) = H(e^{j\hat{\omega}}) e^{j\angle H(e^{j\hat{\omega}})}$$

# PLOT of FREQ RESPONSE

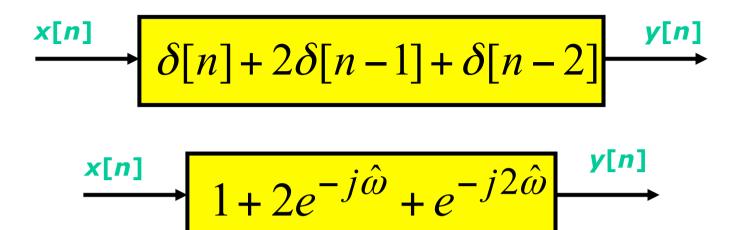
Magnitude of Frequency Response of FIR Filter with Coefficients  $\{b_k\} = \{1,2,1\}$   $\begin{pmatrix} b_k \\ b_k \end{pmatrix} = \{1,2,1\}$ 

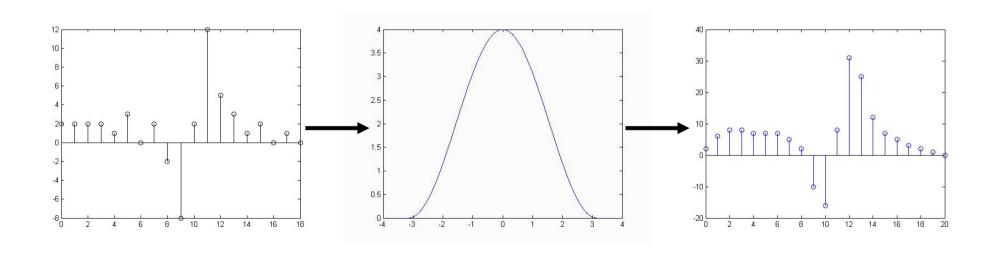


#### FILTER TYPES

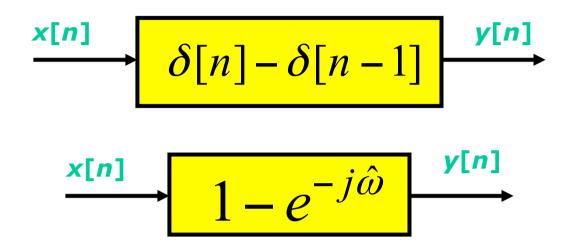
- LOW-PASS FILTER (LPF)
  - BLURRING
  - ATTENUATES HIGH FREQUENCIES
- HIGH-PASS FILTER (HPF)
  - SHARPENING for IMAGES
  - BOOSTS THE HIGHS
  - REMOVES DC
- BAND-PASS FILTER (BPF)

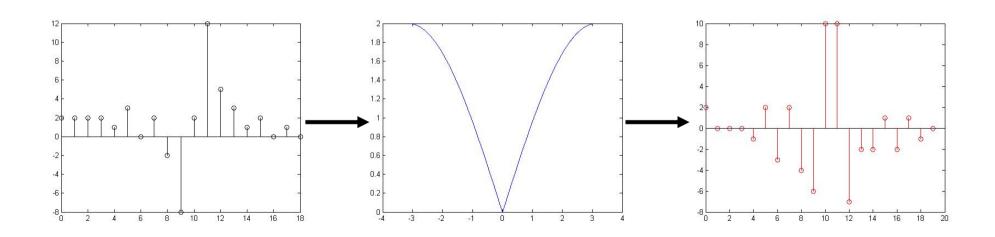
## LOW-PASS FILTER EXAMPLE



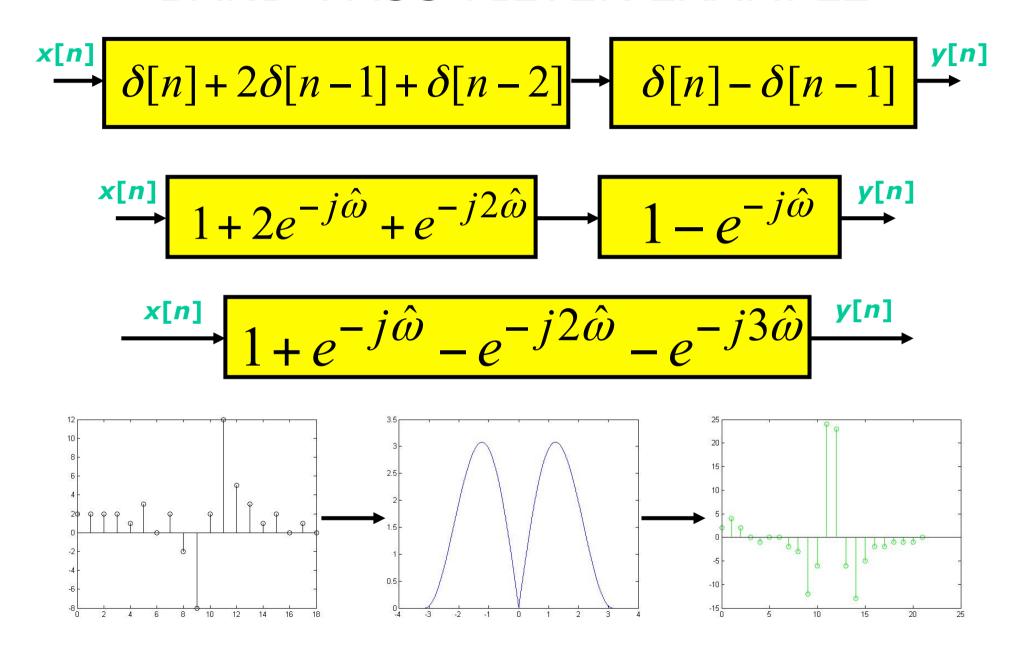


## HIGH-PASS FILTER EXAMPLE

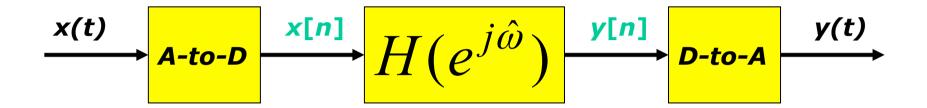




## BAND-PASS FILTER EXAMPLE

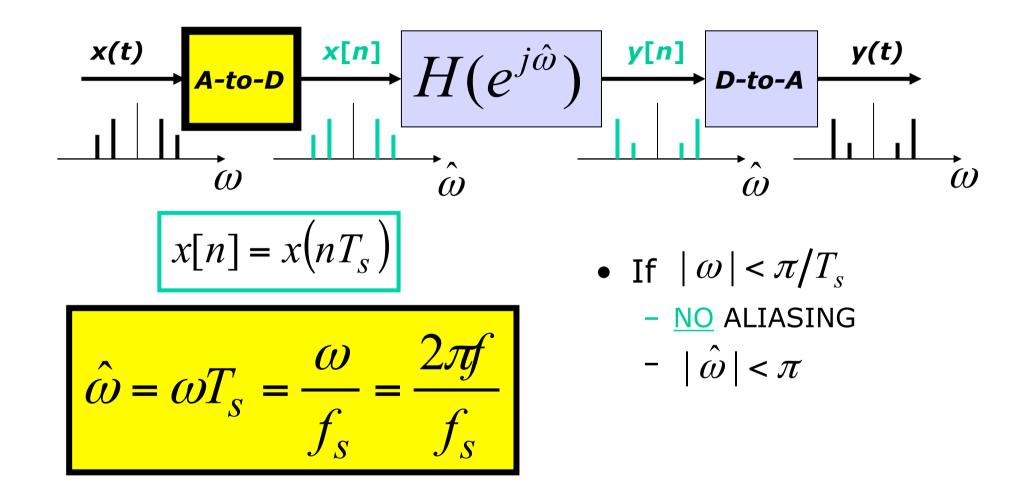


# DIGITAL FILTERING OF ANALOG SIGNALS

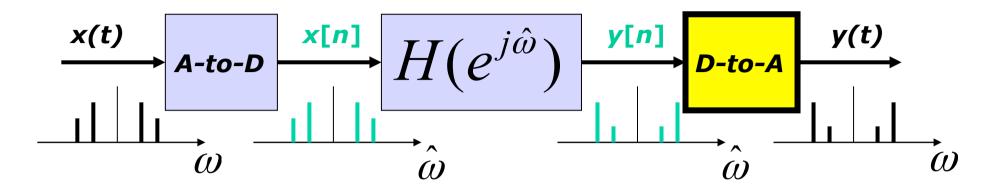


- Use discrete-time filters to filter continuous-time signals that have been sampled
- What is the effect of the filter on the continuoustime input x(t)?
- What is the equivalent analog frequency response?

# FREQUENCY SCALING



# D-A FREQUENCY SCALING



TIME SAMPLING:

$$t = nT_s \Rightarrow n \leftarrow tf_s$$

RECONSTRUCT up to 0.5f<sub>s</sub>
 FREQUENCY SCALING

$$\omega = \hat{\omega} f_s$$

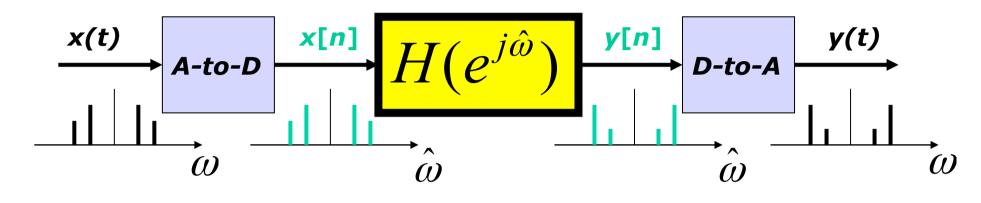
• If input is  $x(t) = Ae^{j\phi}e^{j\omega t}$ ,

output is 
$$y(t) = H(e^{j(\omega T_S)})Ae^{j\phi}e^{j\omega t}$$

ANALOG FREQUENCY RESPONSE

for frequencies  $\omega$  such that  $-\pi/T_{s} < \omega < \pi/T_{s}$ 

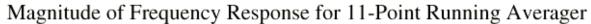
# 11-pt AVERAGER Example

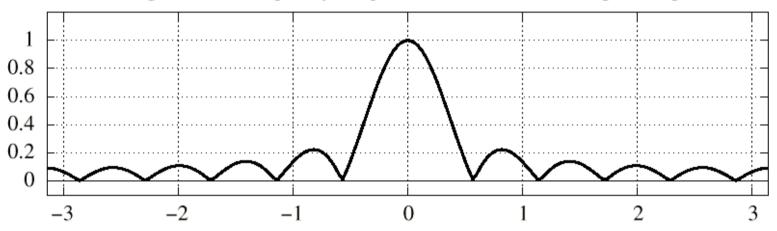


$$y[n] = \sum_{k=0}^{10} \frac{1}{11} x[n-k]$$

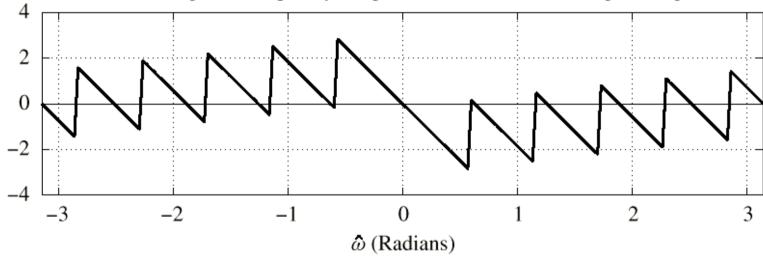
$$H(e^{j\hat{\omega}}) = \frac{\sin(\frac{11}{2}\hat{\omega})}{11\sin(\frac{1}{2}\hat{\omega})}e^{-j5\hat{\omega}}$$

# 11-pt AVERAGER

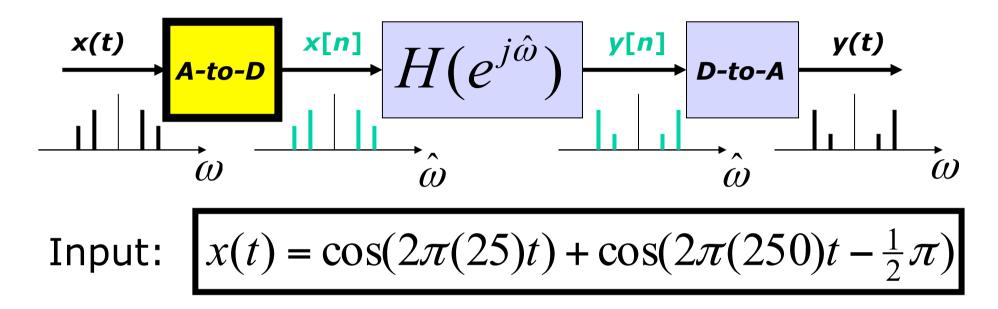




Phase Angle of Frequency Response for 11-Point Running Averager



# 11-pt AVERAGER Example

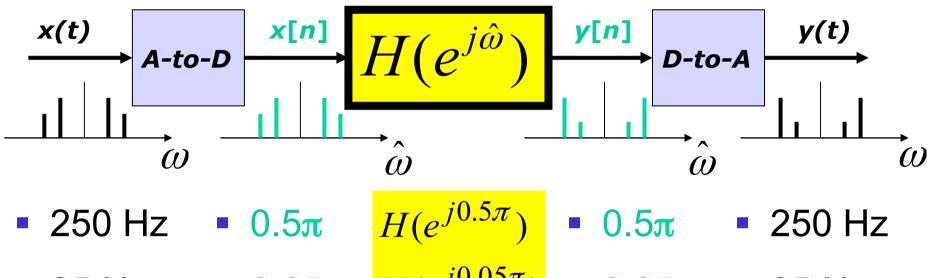


Input frequencies: 25 Hz and 250 Hz

Sampling frequency:  $f_s = 1000 \text{ Hz}$ 

Note:  $f_s > 2 f_{max}$  so no aliasing and x(t) can be reconstructed from x[n]

# TRACK the FREQUENCIES



■ 250 Hz ■ 
$$0.5\pi$$

$$0.5\pi$$

$$H(e^{j0.5\pi})$$

• 25 Hz • 
$$0.05\pi$$
  $H(e^{j0.05\pi})$ 

#### $f_{s} = 1000 \text{ Hz}$

$$x_1(t) = \cos(2\pi(25)t)$$
 ,  $T_s = \frac{1}{1000}$ 

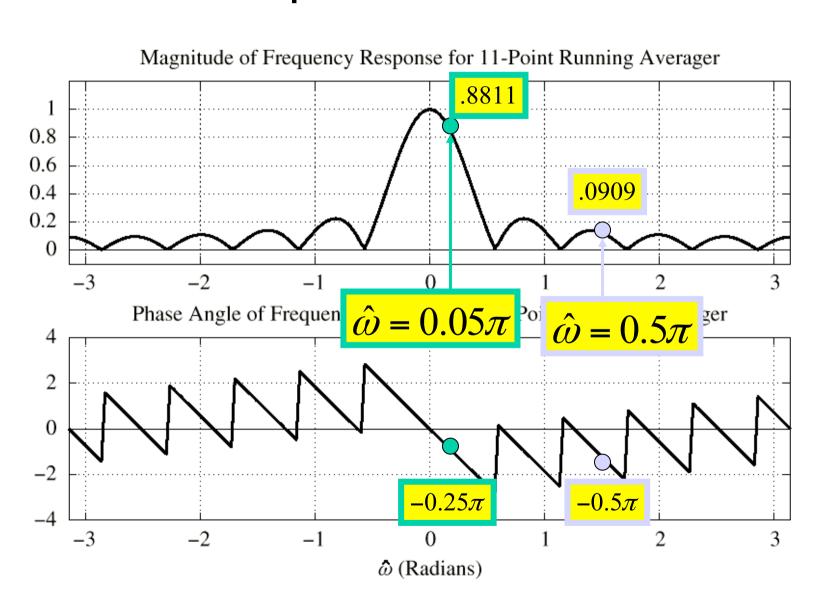
$$\begin{vmatrix} x_1(t) = \cos(2\pi(25)t) & , & T_s = \frac{1}{1000} \\ x_1(nT_s) = \cos\left(\frac{2\pi(25)n}{1000}\right) = \cos\left(\frac{\pi}{20}n\right) \end{vmatrix}$$

#### **NO new freqs**

#### **WARNING**:

When there is aliasing, y(t) will have different frequency components than x(t)

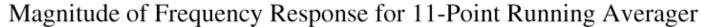
# FREQUENCY RESPONSE OF 11-pt AVERAGER

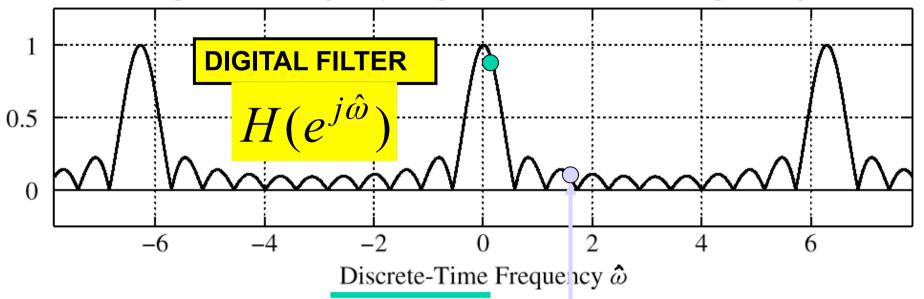


#### **EVALUATE OUTPUT**

Input: 
$$x(t) = \cos(2\pi(25)t) + \cos(2\pi(250)t - \frac{1}{2}\pi)$$

$$H(e^{j0.05\pi}) = 0.881)e^{-j0.25\pi}$$
  $H(e^{j0.5\pi}) = 0.0909e^{-j0.5\pi}$  MAG. SCALE PHASE CHANGE Output:  $y(t) = .8811\cos\left(2\pi(25)t - \frac{\pi}{4}\right) + .0909\cos(2\pi(250)t - \pi)$ 





Equivalent Continuous-Time Frequency Response for  $f_s = 1000$ 

