

# Circuits and Systems I

#### LECTURE #7 Bandlimited Reconstruction



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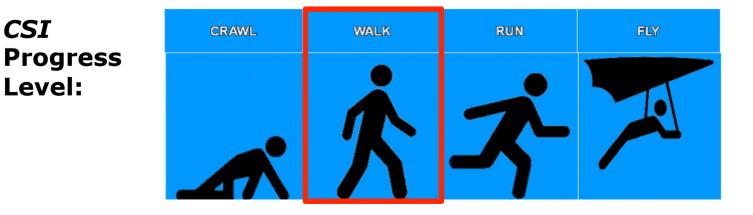
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# Outline - Today

- Today <> Section 4-4 Section 4-5
- Next week <> BONUS EXAM REVIEW!
- Next lecture <> Section 5-1 Section 5-2
  - Section 5-3



# Lecture Objectives

- FOLDING: a type of ALIASING
- DIGITAL-to-ANALOG CONVERSION is
  - Reconstruction from samples
    - SAMPLING THEOREM applies
  - Smooth <u>Interpolation</u>
- Mathematical Model of D-to-A
  - SUM of SHIFTED PULSES
    - Linear Interpolation example

## Signal Types



- Convert x(t) to numbers stored in memory
- D-to-A
  - Convert y[n] back to a "continuous-time" signal, y(t)
  - y[n] is called a "discrete-time" signal

# Sampling x(t)

UNIFORM SAMPLING at t = nT<sub>s</sub>
 IDEAL: x[n] = x(nT<sub>s</sub>)



Shannon Sampling Theorem

A continuous-time signal x(t) with frequencies no higher than  $f_{\text{max}}$  can be reconstructed exactly from its samples  $x[n] = x(nT_s)$ , if the samples are taken at a rate  $f_s = 1/T_s$  that is greater than  $2f_{\text{max}}$ .

# Nyquist Rate

- "Nyquist Rate" Sampling
  - $f_s > \underline{TWICE}$  the HIGHEST Frequency in x(t)
  - "Sampling above the Nyquist rate"

#### • BANDLIMITED SIGNALS

- DEF: x(t) has a HIGHEST FREQUENCY COMPONENT in its SPECTRUM
- NON-BANDLIMITED EXAMPLE
  - TRIANGLE WAVE is NOT BANDLIMITED

# SPECTRUM for x[n]

- INCLUDE ALL SPECTRUM LINES
  - ALIASES
    - ADD INTEGER MULTIPLES of  $2\pi$  and  $-2\pi$
  - FOLDED ALIASES
    - ALIASES of NEGATIVE FREQS
- PLOT versus NORMALIZED FREQUENCY
  - i.e., DIVIDE  $f_o$  by  $f_s$

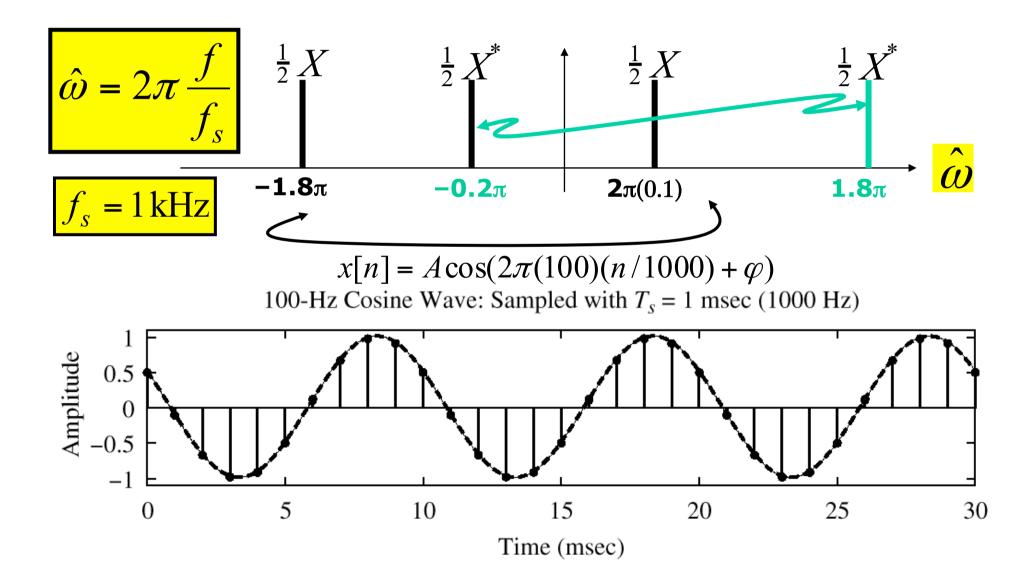
$$\hat{\omega} = 2\pi \frac{f}{f_s} + 2\pi \ell$$

### Example: Spectrum

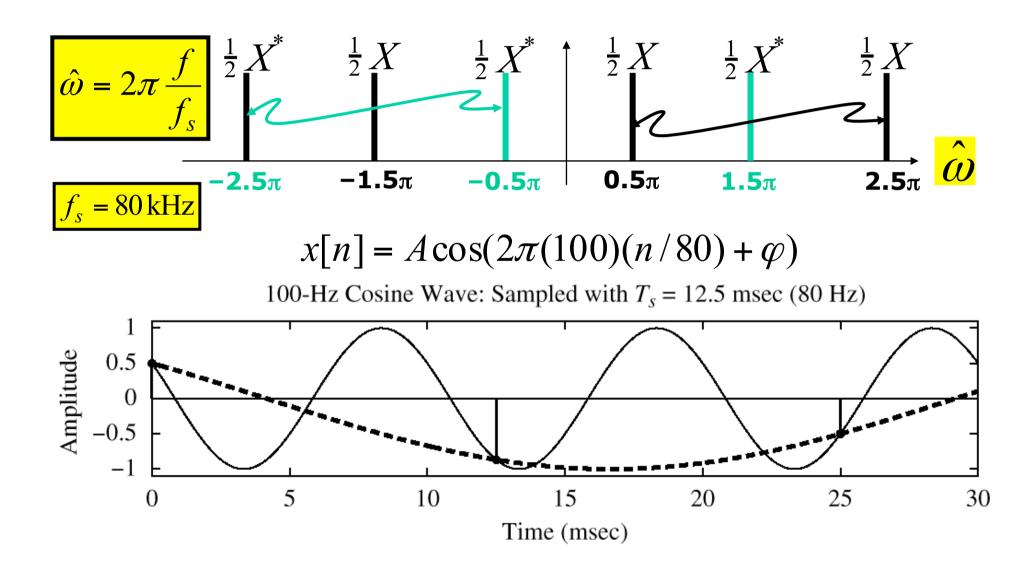
- $x[n] = Acos(0.2\pi n + \phi)$
- FREQS @  $0.2\pi$  and  $-0.2\pi$
- ALIASES:
  - $\{2.2\pi, 4.2\pi, 6.2\pi, ...\}$  &  $\{-1.8\pi, -3.8\pi, ...\}$
  - EX:  $x[n] = Acos(4.2\pi n + \phi)$
- ALIASES of **NEGATIVE** FREQ:

-  $\{1.8\pi, 3.8\pi, 5.8\pi, ...\}$  &  $\{-2.2\pi, -4.2\pi ...\}$ 

## Spectrum (More Lines)



#### Spectrum (Aliasing Case)



## Folding (a type of ALIASING)

EXAMPLE: 3 different x(t); same x[n]

$$f_{s} = 1000$$

$$\cos(2\pi(100)t) \rightarrow \cos[2\pi(0.1)n]$$

$$\hat{\omega} = 2\pi \frac{100}{1000} = 2\pi(0.1)$$

$$\cos(2\pi(1100)t) \rightarrow \cos[2\pi(0.1)n] = \cos[2\pi(0.1)n]$$

$$\cos(2\pi(900)t) \rightarrow \cos[2\pi(0.9)n]$$

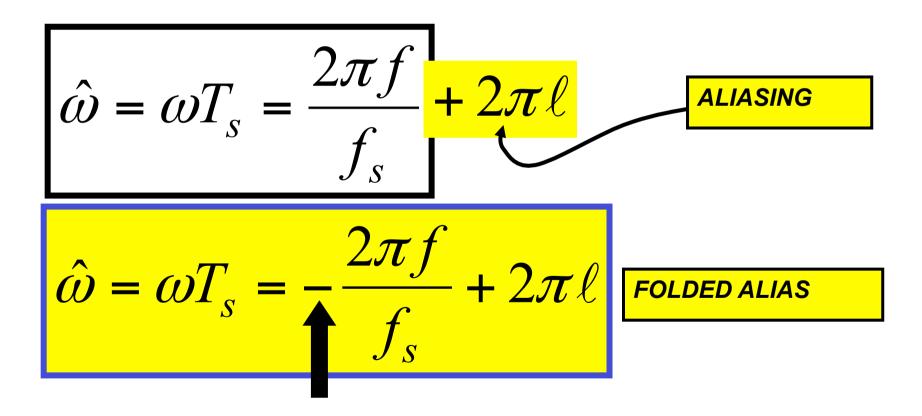
$$= \cos[2\pi(0.9)n - 2\pi n] = \cos[2\pi(-0.1)n] = \cos[2\pi(0.1)n]$$

100

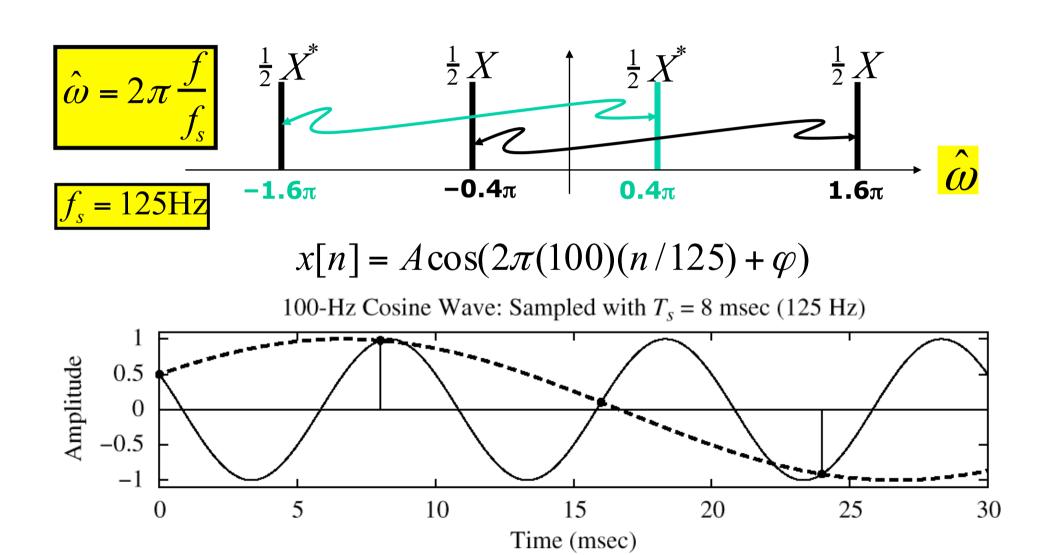
900 Hz "folds" to 100 Hz when f<sub>s</sub>=1kHz

## Digital Frequency $\hat{\omega}$ Again

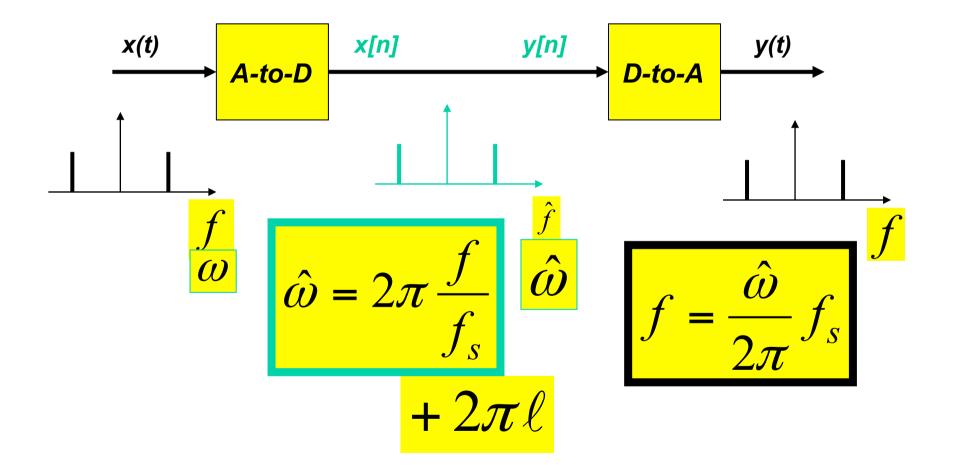
# Normalized Radian Frequency



## Spectrum (Folding Case)



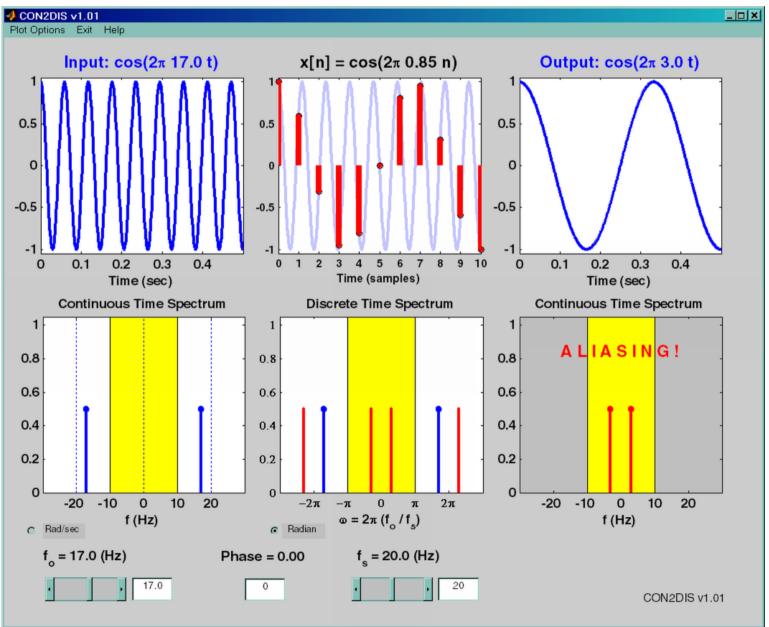
#### **Frequency Domains**



# Demos from Chapter 4

- CD-ROM DEMOS
- SAMPLING DEMO (con2dis GUI)
  - Different Sampling Rates
    - Aliasing of a Sinusoid
- STROBE DEMO
  - Synthetic vs. Real
  - Television SAMPLES at 30 fps in the US / 25 fps in EU
- Sampling & Reconstruction

## SAMPLING GUI (con2dis)



## **D-to-A Reconstruction**

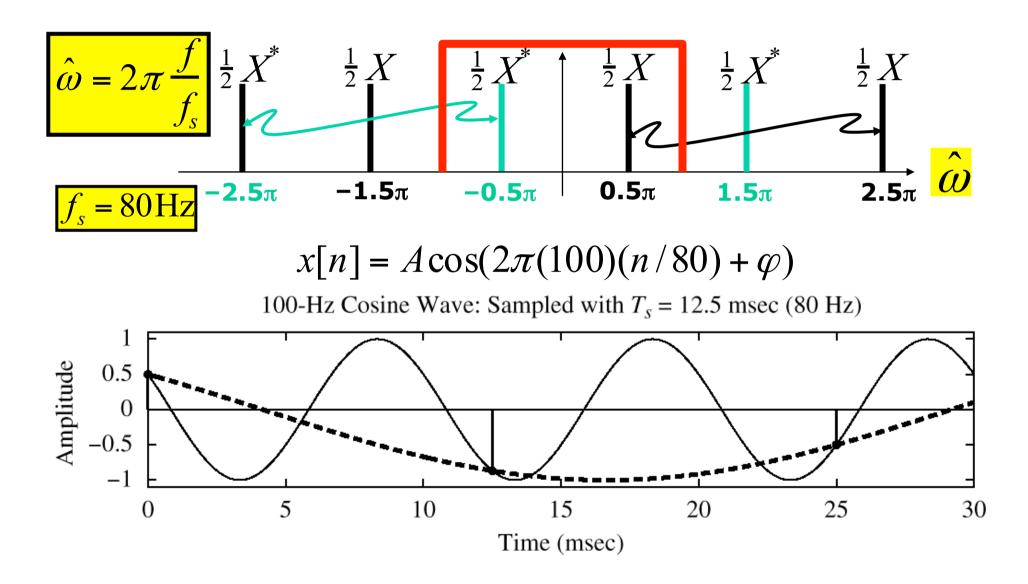


- Create continuous y(t) from y[n]
  - <u>IDEAL</u>
    - If you have formula for y[n]
  - Replace n in y[n] with  $f_st$
  - $y[n] = Acos(0.2\pi n + \phi)$  with  $f_s = 8000$  Hz
  - $y(t) = A\cos(2\pi(800)t + \phi)$

# D-to-A is AMBIGUOUS !

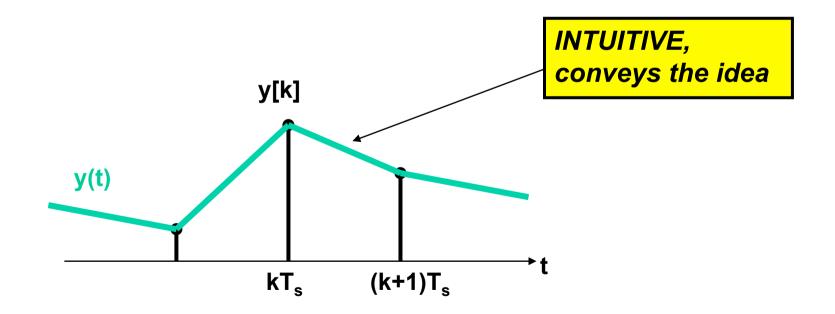
- ALIASING
  - Given y[n], which y(t) do we pick ? ? ?
  - INFINITE NUMBER of y(t)
    - PASSING THRU THE SAMPLES, y[n]
  - D-to-A RECONSTRUCTION MUST CHOOSE ONE OUTPUT
- RECONSTRUCT THE <u>SMOOTHEST</u> ONE
  - THE LOWEST FREQ, if y[n] = sinusoid

### Spectrum (Aliasing Case)



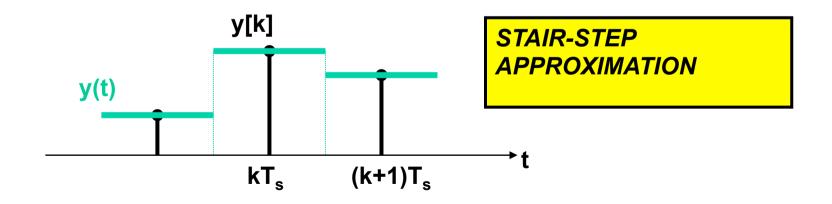
## Reconstruction (D-to-A)

- CONVERT STREAM of NUMBERS to x(t)
- "CONNECT THE DOTS"
- INTERPOLATION

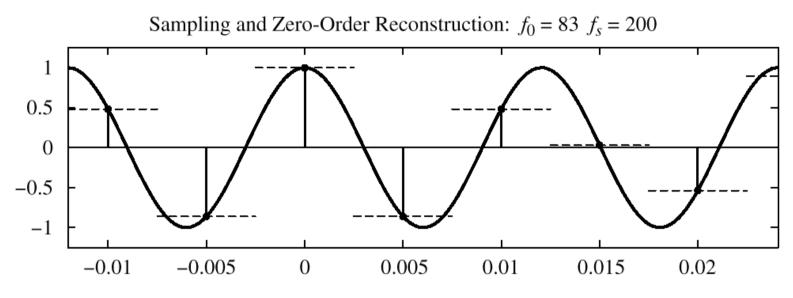


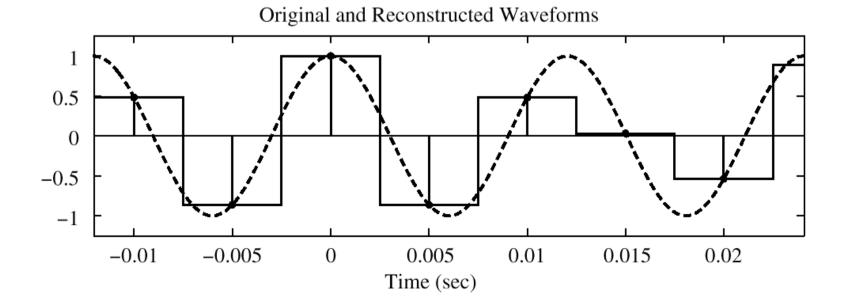
## Sample and Hold Device

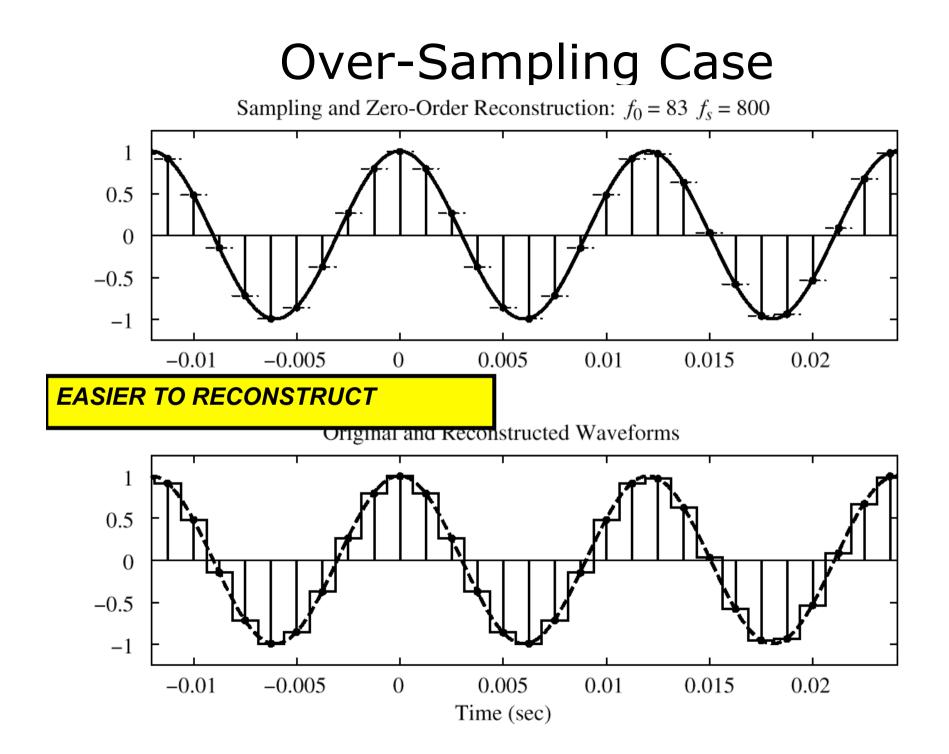
- CONVERT y[n] to y(t)
  - y[k] should be the value of y(t) at  $t = kT_s$
  - Make y(t) equal to y[k] for
    - kT<sub>s</sub> -0.5T<sub>s</sub> < t < kT<sub>s</sub> +0.5T<sub>s</sub>



#### Square Pulse Case







### Mathematical Model for D-to-A

$$y(t) = \sum_{n=-\infty}^{\infty} y[n]p(t - nT_s)$$

SQUARE PULSE:

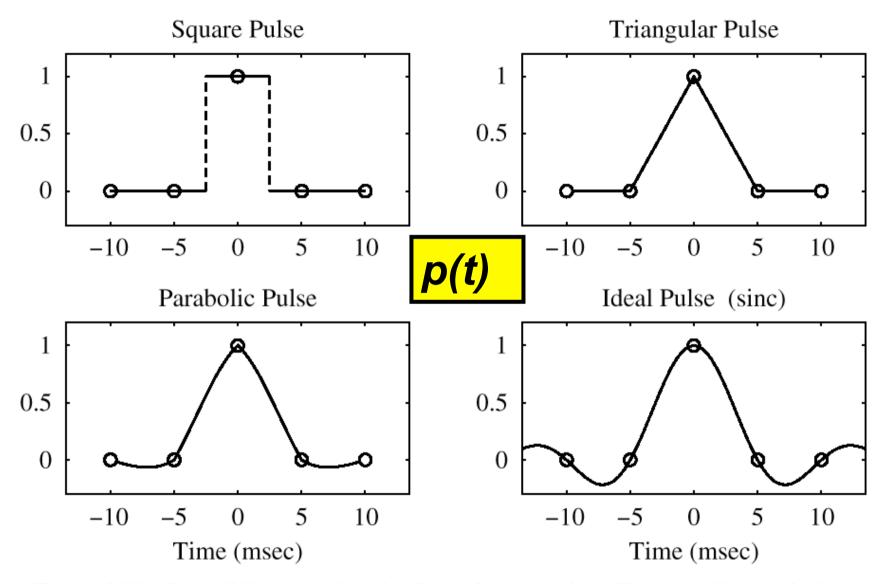
$$p(t) = \begin{cases} 1 & -\frac{1}{2}T_s < t \le \frac{1}{2}T_s \\ 0 & \text{otherwise} \end{cases}$$

## Expand the Summation

$$\sum_{n=-\infty}^{\infty} y[n]p(t-nT_s) =$$

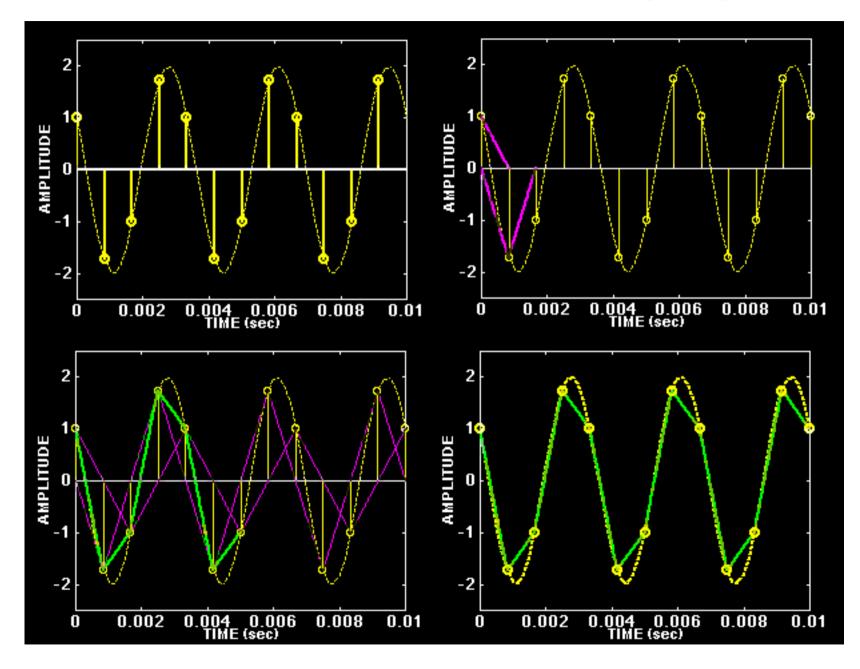
$$\mathbf{K} + y[0]p(t) + y[1]p(t - T_s) + y[2]p(t - 2T_s) + \mathbf{K}$$

- SUM of SHIFTED PULSES p(t-nT<sub>s</sub>)
  - "WEIGHTED" by y[n]
  - CENTERED at  $t=nT_s$
  - SPACED by  $\mathrm{T}_{\mathrm{s}}$ 
    - RESTORES "REAL TIME"



**Figure 4.17** Four different pulses for D-to-C conversion. The sampling period is  $T_s = 0.005$ , i.e.,  $f_s = 200$  Hz. Note that the duration of each pulse is approximately one or two times  $T_s$ .

#### TRIANGULAR PULSE (2X)



## **Optimal Pulse?**

